The concepts of self-timed circuits date back to the early days of computers, but the increased difficulty of building such systems has limited their application in digital design. Self-timed concepts, however, are used extensively in advanced dynamic memory components. The best place to start in finding out more about self-timed circuits is in Seitz's chapter in Mead and Conway's book referenced above. A research group, led by Professor Alain Martin of the California Institute of Technology, has succeeded in implementing a complete 32-bit microprocessor using self-timed techniques. Their work is reported in the 10th CALTECH Conference VLSI Proceedings; the conference was held in March 1989 in Pasadena, CA.

Exercises

- **6.1** (Simple Circuits with Feedback) Build a feedback circuit with cross-coupled NAND gates. What input conditions cause the state of this latch-like device to be reset? To be set? Does this circuit have forbidden inputs? If so, what are they?
- 6.2 (Simple Circuits with Feedback) An R-S latch can be used to determine which of two events has occurred first. Design a circuit with three inputs and three outputs that determines which of three single pole/single throw switches connected to the inputs has been opened first. The circuit will produce a logic 1 on the output that corresponds to that input. Discuss how you would expand this circuit to a larger number of inputs, say 12 inputs or 30 inputs.
- 6.3 (Setup and Hold Times) Imagine that it is possible to have storage devices with negative setup and hold times. What do you think such a concept would be? Draw timing diagrams to illustrate your answer.
- 6.4 (D Flip-Flop) Add preset and clear inputs to the edge-triggered D flip-flop of Figure 6.24. Draw the logic schematic of the revised circuit.
- 6.5 (D Flip-Flop) How would you implement a negative edge-triggered D flip-flop using NAND gates only? What changes are necessary to make this a positive edge-triggered device?
- 6.6 (J-K Flip-Flop) How would you implement a J-K master/slave flip-flop, such as the circuit in Figure 6.22, using NAND gates only? Assume the master and slave latches are actually controlled by the signals \overline{R} and \overline{S} .
- 6.7 (*J-K Flip-Flop*) Starting with the basic circuit schematic for the master/slave *J-K* flip-flop, show how to add asynchronous preset and clear inputs to force the flip-flop into a 1 (preset) or 0 (clear)

- state. Draw a timing waveform for the preset input, clear input, clock, master stage outputs (P, \overline{P}) , and slave stage outputs (Q, \overline{Q}) showing the operation of preset and clear.
- 6.8 (J-K Flip-Flop) J-K master/slave flip-flops exhibit the phenomenon of ones catching. Briefly explain why this takes place. Can a master/slave flip-flop catch 0's? Explain why or why not.
- **6.9** (Flip-Flops) The basic functionality of a D flip-flop can be implemented by a J-K flip-flop simply by connecting the input signal D to the J-K flip-flop's J input and \overline{D} to the K input.
 - a. Show that this is true by comparing the characteristic equations for a D flip-flop and a J-K flip-flop.
 - **b.** Draw a timing waveform for the clock, input D, and outputs $Q_{\rm pos}$, $Q_{\rm neg}$, and $Q_{\rm ms}$ that illustrates the differences in input/output behavior of a positive edge-triggered D flip-flop, negative edge-triggered D flip-flop, and master/slave D flip-flop (implemented from a J-K master/slave flip-flop as described in Figure 6.22). Include some transitions on D while the clock is asserted.
- **6.10** (Flip-Flops) Given the input and clock transitions given in Figure Ex6.10, draw a waveform for the output of a J-K device, assuming:
 - a. It is a master/slave flip-flop.
 - b. It is a positive edge-triggered flip-flop.
 - ${f c.}$ It is a negative edge-triggered flip-flop. You may assume 0 setup, hold, and propagation times, and that the initial state of the flip-flop is 0.

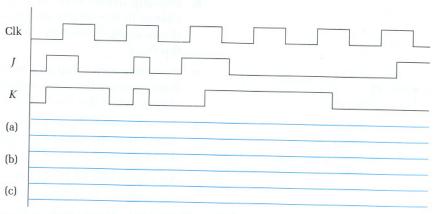


Figure Ex6.10 Timing diagram for Exercise 6.10.

- **6.11** (*Flip-Flops*) Given the input and clock transitions in Figure Ex6.11, indicate the output of a *D* device assuming:
 - a. It is a negative edge-triggered flip-flop.
 - b. It is a master/slave flip-flop.
 - c. It is a positive edge-triggered flip-flop.
 - d. It is a clocked latch. You may assume 0 setup, hold, and propagation times.

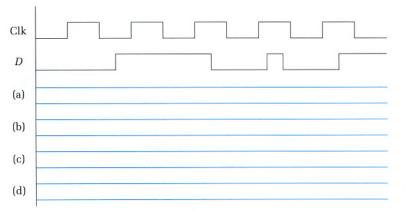
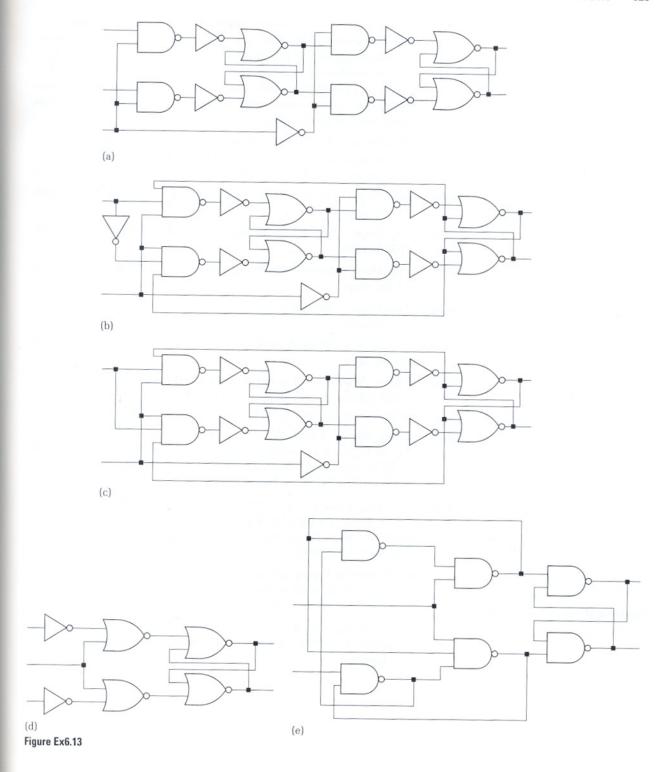


Figure Ex6.11 Timing diagram for Exercise 6.11.

- **6.12** (Flip-Flops) Identify the following statements as either true or false:
 - a. The inputs to a level-sensitive latch always affect its outputs.
 - b. Flip-flop delays from the change in the clock edge to the change in the output are typically shorter than flip-flop hold times, so shift registers can be constructed from cascaded flip-flops.
 - c. Assuming zero setup and hold times, clocked latches and flipflops produce the same outputs as long as the inputs do not change while the clock is asserted.
 - d. A master/slave flip-flop behaves similarly to a clocked latch, except that its output can change only near the rising edge of the clock.
 - **e.** An edge-triggered *D* flip-flop requires more internal gates than a similar device constructed from a *J-K* master/slave flip-flop.
- **6.13** (Flip-Flops) Match each of the following five circuits (Figure Ex6.13) with the phrase that best describes it from the list: (1) clocked R-S latch, (2) clocked D latch, (3) master/slave R-S



- **6.14** (Flip-Flops) Any flip-flop type can be implemented from another type with suitable logic applied to the latter's inputs. Show how to implement a J-K flip-flop starting with a D flip-flop.
- **6.15** (Flip-Flops) Show how to implement a J-K flip-flop starting with a T flip-flop.
- **6.16** (Flip-Flops) Show how to implement a D flip-flop starting with a J-K flip-flop.
- **6.17** (Flip-Flops) Show how to implement a D flip-flop starting with a T flip-flop.
- **6.18** (*Flip-Flops*) Show how to implement a *T* flip-flop starting with a *J-K* flip-flop.
- **6.19** (*Flip-Flops*) Show how to implement a *T* flip-flop starting with a *D* flip-flop.
- 6.20 (Clock Generator) Extend the circuit of Figure 6.37 to generate a three-phase nonoverlapping clock. How would the circuit be extended to generate a four-phase nonoverlapping clock?
- 6.21 (Clock Skew) Given the timing specification of the 74LS74 flipflop of Figure 6.14, what is the worst-case skew in the clock that could be tolerated when one 74LS74 needs to pass its value to another 74LS74, as in Figure 6.29?
- 6.22 (Clocking Issues) Given the sequential logic circuit of Figure Ex6.22, where the flip-flops have worst-case setup times of 20 ns, propagation delays of 13 ns, and hold times of 5 ns, answer the following questions:
 - a. Assuming 0 propagation delay through the combinational logic block, what is the maximum allowable frequency of the clock that controls this subsystem?
 - b. Assuming a typical combinational logic delay of 75 ns and a worst-case delay of 100 ns, how does your answer to part (a) change?

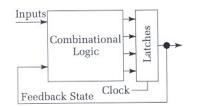


Figure Ex6.22 Sequential circuit for Exercise 6.22.

6.23 (Two-Phase Clocking) Consider the two-phased clocked sequential circuit of Figure Ex6.23. Assume that the ϕ_1 and ϕ_2 latches have 5-ns setup times, 5-ns hold times, and 10-ns propagation delays in the worst case. Given that the combinational logic block has a maximum delay of 100 ns and a typical delay of 75 ns, what is the shortest possible period between the rising edge of ϕ_1 and the falling edge of ϕ_2 ? Show how you obtained your result.

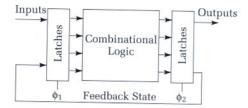


Figure Ex6.23 Two-phase clocked sequential circuit for Exercise 6.23.

- 6.24 (Metastability) You have designed a high-performance disk drive interface. The interface has an internal clock rate of 25 MHz, and asynchronous commands from a computer with a different clock are presented every 200 ns. It works fine, but every few days or weeks it has random operational failures that cause loss of data. There are no component failures, software bugs, or power glitches, and the errors occur mainly for customers who use the interface heavily. Suggest a possible cause of these failures and how you could change the design to reduce the failure rate.
- 6.25 (Metastability) One way to reduce the probability of synchronizer failure is to place two synchronizer flip-flops in series between the asynchronous input and the rest of the synchronous digital system. Why do you think this reduces the problem of metastability?
- **6.26** (Delay Insensitive Handshaking) Draw a simple flowchart for the master side and the slave side algorithms of the four-cycle handshake. Repeat for the two-cycle handshake. How does the complexity of the two approaches compare?
- **6.27** (Practical Matters) How would the debounce circuit of Figure 6.56 change if you were to use an \overline{R} - \overline{S} latch instead of an R-S latch?
- **6.28** (Practical Matters) Suppose you are to design a debouncing circuit using a single pole/single throw (SPST) switch. Can it be done? What problems do you face?

- 6.29 (Practical Matters) Given the discussion of the 555 timer chip, is it possible to obtain duty cycles of less than 50%? If it is, give an example of resistor ratios needed to obtain such a duty cycle. If not, explain why and give a possible scheme for deriving a waveform with less than a 50% duty cycle from the 555's output.
- 6.30 (Practical Matters) Use the equations given in Section 6.6.2 and the chart in Figure 6.58 to configure the 555 timer according to the following specifications. Show all of your intermediate work, coming as close to the specification as you can:
 - a. Choose resistors and a capacitor to obtain a 100-kHz clock with 67% duty cycle.
 - b. Choose resistors to obtain a 1 MHz clock with 75% duty cycle using a 0.1-µF capacitor.
 - c. Choose resistors and a capacitor to obtain a clock period of 140 μs and 75% duty cycle.
 - d. Given a clock period of 1 μ s, $R_a = 9 k\Omega$, $R_b = 1 k\Omega$, what is the duty cycle and what size of capacitor should be used?